## **MP2662**

# 500mA Single-Cell Li-ion Battery Charger with Power Path Management, 1mA Termination and <1µA Battery Leakage

#### DESCRIPTION

The MP2662 is a highly integrated, single-cell, Li-ion/Li-polymer battery charger with system power path management for space-limited portable applications. The MP2662 takes input power from either an AC adapter or a USB port to supply the system load and charge the battery simultaneously. The charger function features pre-charge (PRE.C), constant current fast charge(CC) and constant voltage (CV) regulation, charge termination, and autorecharge.

The power path management function ensures continuous power to the system by automatically selecting the input, battery, or both to power the system. This function features a low dropout regulator from the input to the system and a  $100m\Omega$  switch from the battery to the system. Power path management separates the charging current from the system load, which allows for proper charge termination and keeps the battery in full-charge mode.

The MP2662 provides a system short-circuit protection (SCP) function by limiting the current from the input to the system and the battery to the system. This feature is especially critical for preventing the Li-ion battery from being damaged due to an excessively high current. An on-chip battery under-voltage lockout (UVLO) cuts off the path between the battery and the system if the battery voltage drops below a programmable battery UVLO threshold. This prevents the Li-ion battery from being over-discharged. An integrated I2C control interface allows the MP2662 to program the charging parameters, such as input current limit, input minimum voltage regulation, charging current, battery regulation voltage, safety timer, and battery UVLO.

The MP2662 is available in a 9-pin WLCSP (1.75mmx1.75mm) package.

#### **FEATURES**

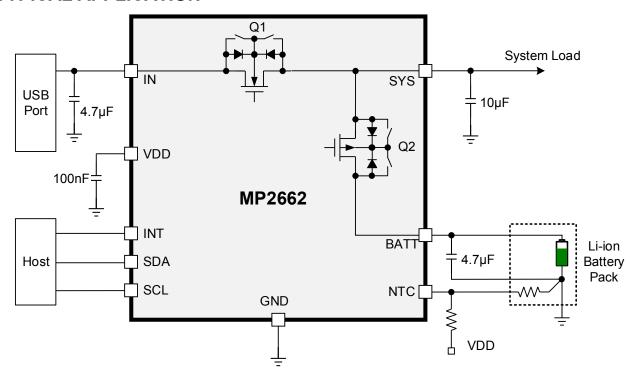
- Fully Autonomous Charger for Single-Cell Li-Ion/Polymer Batteries
- 21V Maximum Input Voltage Rating with Over-Voltage Protection (OVP)
- ±0.5% Charging Voltage Accuracy
- I<sup>2</sup>C Interface for Setting Charging Parameters and Status Reporting
- Fully Integrated Power Switches and No External Blocking Diode Required
- Built-In Robust Charging Protection Including Battery Temperature Monitoring and Programmable Timer
- PCB Over-Temperature Protection (PCB OTP)
- System Reset Function
- Built-In Battery Disconnection Function for Shipping Mode
- Thermal Limiting Regulation On-Chip
- Available in a WLCSP-9 (1.75mmx1.75mm) Package
- Safety-Related Certification:
  - o IEC 62368-1 CB Certification

#### **APPLICATIONS**

- Wearable Devices
- Smart Handheld Devices
- Fitness Accessories
- Smart Watches

All MPS parts are lead-free, halogen-free, and adhere to the RoHS directive. For MPS green status, please visit the MPS website under Quality Assurance. "MPS", the MPS logo, and "Simple, Easy Solutions" are trademarks of Monolithic Power Systems, Inc. or its subsidiaries.

## **TYPICAL APPLICATION**



#### ORDERING INFORMATION

Part Number*	Package	Top Marking
MP2662GC-xxxx**	WLCSP-9 (1.75mmx1.75mm)	See Below
EVKT-MP2662	Evaluation Kit	

<sup>\*</sup>For Tape & Reel, add suffix -Z (e.g. MP2662GC-xxxx-Z).

#### **TOP MARKING**

**JAY** 

LLL

JA: Product code of MP2662GC

Y: Year code LLL: Lot number

#### **EVALUATION KIT EVKT-MP2662**

EVKT-MP2662 kit contents: (Items below can be ordered separately)

#	Part Number	Item	Quantity
1	EV2662-C-01A	MP2662 evaluation board	1
2	EVKT-USBI2C-02-bag	Includes one USB to $I^2C$ communication interface, one USB cable, and one ribbon cable	1
3	Online resources	Include datasheet, user guide, product brief, and GUI	1

#### Order direct from MonolithicPower.com or our distributors.

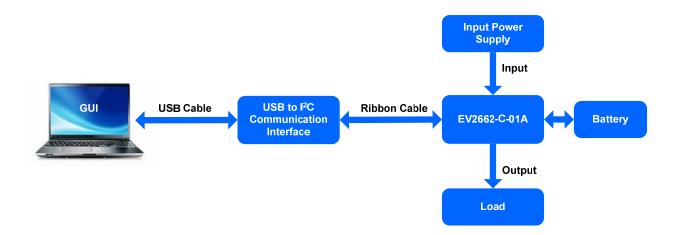
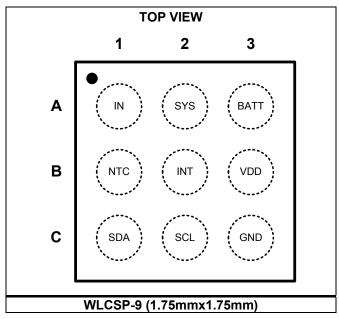


Figure 1: EVKT-MP2662 Evaluation Kit Set-Up

<sup>\*\*&</sup>quot;xxxx" is the register setting option. The factory default is "0000." This content can be viewed in the I<sup>2</sup>C register map. Please contact an MPS FAE to obtain an "xxxx" value.

## **PACKAGE REFERENCE**



## **PIN FUNCTIONS**

Pin#	Name	I/O	Description
A1	IN	Power	<b>Input power pin.</b> Place a ceramic capacitor from IN to GND, and as close to the IC as possible.
A2	SYS	Power	<b>System power supply.</b> Place a ceramic capacitor from SYS to GND, and as close to the IC as possible.
А3	BATT	Power	<b>Battery pin.</b> Place a ceramic capacitor from BATT to GND as close to the IC as possible.
B1	NTC	I	<b>Temperature sense input.</b> Connect a negative temperature coefficient thermistor to NTC. Configure the hot and cold temperature window by placing a resistor divider from VDD to NTC to GND. Charging is suspended when NTC is out of its range. Pull NTC to VDD if the NTC function is not used.
B2	INT	I/O	Interrupt signal. INT can send a charging status and fault interrupt signal to the host. INT is also used to disconnect the system from the battery. Pull INT from high to low for longer than $t_{RST\_DGL}$ (16s by default). The battery FET turns off and turns on again automatically after $t_{RST\_DUR}$ (4s by default), regardless of the INT state. Both $t_{RST\_DGL}$ and $t_{RST\_DUR}$ can be configured via the $l^2C$ interface.
В3	VDD	Power	Internal control power supply pin. Connect a 100nF ceramic capacitor from VDD to GND. No external load is allowed.
C1	SDA	I/O	I <sup>2</sup> C interface data. Connect SDA to the logic rail through a 10kΩ resistor.
C2	SCL	I	<b>I<sup>2</sup>C interface clock.</b> Connect SCL to the logic rail through a 10kΩ resistor.
C3	GND	Power	Ground.

ABSOLUTE MAXIMUM	I RATINGS (1)
IN	0.3V to +21V
SYS0.3V to +5.3V	V (5.5V for 500µs)
All other pins to GND	0.3V to +6V
Continuous power dissipation	
	Ó.88W
Junction temperature	
Lead temperature (solder)	260°C
Storage temperature	
Recommended Operating	Conditions (3)
Supply voltage (V <sub>IN</sub> ) 4.35V to	
I <sub>IN</sub>	Up to 500mA
IDSCHG	Up to 3.2A (5)
I <sub>CHG</sub>	
V <sub>BATT_REG</sub>	Up to 4.545V
Operating junction temp. (T <sub>J</sub> )	

Thermal Resistance (4)  $\theta_{JA}$   $\theta_{JC}$ WLCSP-9 (1.75mmx1.75mm) ... 114 ... 12 .. °C/W

#### NOTES:

- Exceeding these ratings may damage the device.
   The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub> (MAX), the junction-toambient thermal resistance  $\theta_{JA}$ , and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by  $P_D$  (MAX) =  $(T_J)$ (MAX)- $T_A)/\theta_{JA}$ . Exceeding the maximum allowable power dissipation produces an excessive die temperature, causing the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent
- damage.

  3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.
- 5) Guaranteed by design.

## **ELECTRICAL CHARACTERISTICS**

 $V_{IN}$  = 5.0V,  $V_{BATT}$  = 3.5V,  $T_A$  = 25°C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Input Source and Batter	y Protection				<u> </u>	
Input under-voltage lockout threshold	V <sub>IN_UVLO</sub>	Input falling	3.63	3.73	3.83	V
Input under-voltage lockout threshold hysteresis		Input rising		170		mV
Input over-voltage protection threshold	V <sub>IN_OVP</sub>	Input rising threshold	5.85	6	6.15	V
Input over-voltage protection threshold hysteresis				350		mV
Input vs. battery voltage headroom threshold	V <sub>HDRM</sub>	Input rising vs. battery	80	130	170	mV
Input vs. battery voltage headroom threshold hysteresis				90		mV
BATT input voltage (5)	V <sub>BATT</sub>				4.5	V
		BATT voltage falling, REG01h, bits[2:0] = 000	2.3	2.4	2.5	
Battery under-voltage lockout threshold	VBATT_UVLO	BATT voltage falling, REG01h, bits[2:0] = 100	2.66	2.76	2.86	V
		BATT voltage falling, REG01h, bits[2:0] = 111	2.93	3.03	3.13	
Battery under-voltage threshold hysteresis		V <sub>BATT_UVLO</sub> = 2.76V		190		mV
Battery over-voltage protection threshold	V <sub>BATT_OVP</sub>	Rising, higher than VBATT_REG		130		mV
Battery over-voltage protection hysteresis				60		1110
Power Path Managemen	t					
		$V_{IN} = 5.5V$ , $R_{SYS} = 100\Omega$ , $I_{CHG} = 0A$ , REG07h, bits[3:0] = 0000, $V_{SYS\_REG} = 4.2V$	-2		2	%
Regulated system output voltage accuracy	Vsys_reg_ acc	$V_{IN}$ = 5.5V, $R_{SYS}$ = 100 $\Omega$ , $I_{CHG}$ = 0A, REG07h, bits[3:0] = 1001, $V_{SYS\_REG}$ = 4.65V	-2		2	%
		$V_{IN}$ = 5.5V, $R_{SYS}$ = 100 $\Omega$ , $I_{CHG}$ = 0A, REG07h, bits[3:0] = 1111, $V_{SYS\_REG}$ = 4.95V	-2		2	%
		REG00h, bits[3:0] = 0000, I <sub>IN_LIM</sub> = 50mA	30	40	50	
Input current limit	liki i sa	REG00h, bits[3:0] = 0011, I <sub>IN_LIM</sub> = 140mA	112	126	140	mA
mpat carrent iiinit	lin_lim	REG00h, bits[3:0] = 1001, I <sub>IN_LIM</sub> = 320mA	275	300	325	IIIA
		REG00h, bits[3:0] = 1111, $I_{IN\_LIM}$ = 500mA	440	470	500	
Input minimum voltage		REG00h, bits[7:4] = 0000, V <sub>IN_MIN</sub> = 3.88V	3.68	3.88	4.18	
Input minimum voltage regulation	V <sub>IN_MIN</sub>	REG00h, bits[7:4] = 1001, V <sub>IN_MIN</sub> = 4.60V	4.40	4.60	4.75	V
- J		REG00h, bits[7:4] = 1111, V <sub>IN_MIN</sub> = 5.08V	4.88	5.08	5.35	

## **ELECTRICAL CHARACTERISTICS** (continued) $V_{IN} = 5.0V$ , $V_{BATT} = 3.5V$ , $T_A = 25$ °C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units	
IN to SYS switch on resistance	R <sub>ON_Q1</sub>	V <sub>IN</sub> = 4.5V, I <sub>SYS</sub> = 100mA		290		mΩ	
Input quiescent current	lu. a	$V_{IN}$ = 5.5V, EN_HIZ = 0, CEB = 0, charge enable, I <sub>CHG</sub> = 0A, I <sub>SYS</sub> = 0A		1.9		mΛ	
input quiescent current	lin_q	$V_{IN}$ = 5.5V, EN_HIZ = 0, CEB = 1, charge disable		1.7		- mA	
		$V_{IN}$ = 5V, CEB = 0, $I_{SYS}$ = 0A, charge done, $V_{BATT}$ = 4.35V		43			
		V <sub>IN</sub> = GND, CEB = 1, I <sub>SYS</sub> = 0A, V <sub>BATT</sub> = 4.35V, disable PCB_OTP function, not including the current from the external NTC resistor		6.5	7.5		
Battery quiescent current	IBATT_Q	$V_{\text{IN}}$ = GND, CEB = 1, $I_{\text{SYS}}$ = 0A, $V_{\text{BATT}}$ = 4.35V, enable PCB_OTP function, not including the current from the external NTC resistor		14.5	21	μA	
		$V_{\text{IN}}$ = GND, CEB = 1, $I_{\text{SYS}}$ = 0A, $V_{\text{BATT}}$ = 4.35V, enable PCB_OTP function, not including the current from the external NTC resistor, enable watchdog		22.5			
		V <sub>BATT</sub> = 4.5V, V <sub>IN</sub> = V <sub>SYS</sub> = GND, FET_DIS = 1, shipping mode			350	nA	
Battery FET on resistance	RON_Q2	V <sub>IN</sub> < 2V, V <sub>BATT</sub> = 3.5V, I <sub>SYS</sub> = 100mA		100		mΩ	
Battery FET discharge	1	REG03h, bits[7:4] = 0001, I <sub>DSCHG</sub> = 400mA	370	490	585	mA	
current limit	IDSCHG	REG03h, bits[7:4] = 1001, I <sub>DSCHG</sub> = 2000mA		2400(5)			
SYS reverse to BATT switch leakage		$V_{SYS}$ = 4.65V, $V_{IN}$ = 5V, $V_{BATT}$ = GND, EN_HIZ = 1, CEB = 1, charge disable			100	nA	
Ideal diode forward voltage in supplement mode	V <sub>FWD</sub>	50mA discharge current		30		mV	
Shipping Mode							
Enter shipping mode deglitch time	tsmen_dgl	REG06h, bit[5] is set from 0 to 1, REG09h, bits[7:6] = 00		1		s	
Exit shipping mode by INT or $V_{\text{IN}}$ plug-in	t <sub>SMEX_DGL</sub>	INT is pulled low		2		s	
Auto-Reset Mode							
Decet by INIT	1	REG01h, bits[7:6] = 00		8			
Reset by INT	<b>t</b> RST_DGL	REG01h, bits[7:6] = 10		16		S	
Battery FET off lasting time	tpor sus	REG01h, bit[5] = 0		2		s	
Dattery I LT on lasting time	t <sub>RST_DUR</sub>	REG01h, bit[5] = 1		4		3	
	1			1		_	

## **ELECTRICAL CHARACTERISTICS** (continued) $V_{IN} = 5.0V$ , $V_{BATT} = 3.5V$ , $T_A = 25$ °C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Battery Charger						
		REG04h, bits[7:2] = 000000, V <sub>BATT_REG</sub> = 3.6V	3.582	3.6	3.618	
Battery charge voltage regulation	V	REG04h, bits[7:2] = 101000, V <sub>BATT_REG</sub> = 4.2V	4.179	4.2	4.221	V
	V <sub>BATT_REG</sub>	REG04h, bits[7:2] = 110010, VBATT_REG = 4.38V	4.358	4.38	4.4	v
		REG04h, bits[7:2] = 111110, V <sub>BATT_REG</sub> = 4.53V	4.522	4.53	4.568	
		REG02h, bits[5:0] = 000000, I <sub>CC</sub> = 8mA	6.9	8	8.5	
		REG02h, bits[5:0] = 001011, I <sub>CC</sub> = 96mA	89	96	103	mA
Fast charge current	Icc	REG02h, bits[5:0] = 100000, Icc = 264mA	251	264	285	
		REG02h, bits[5:0] = 111000, Icc = 456mA	420	456	484	
Junction temperature regulation <sup>(5)</sup>	T <sub>J_REG</sub>	Thermal_Limit = 120°C		120		°C
Pre-charge current	I <sub>PRE</sub>	I <sub>PRE</sub> = I <sub>TERM</sub>	1		31	mA
	I <sub>TERM</sub>	REG03h, bits[3:0] = 0000, I <sub>TERM</sub> = 1mA	8.0	0.93	1.05	_
Charge termination current		REG03h, bits[3:0] = 0001, I <sub>TERM</sub> = 3mA	2.7	3	3.3	mA
threshold		REG03h, bits[3:0] = 0101, I <sub>TERM</sub> = 11mA	10	11	12	
		REG03h, bits[3:0] = 0101, I <sub>TERM</sub> = 31mA	28	31	34	
Termination deglitch time	tterm_dgl			3.2		S
Pre-charge to fast charge threshold	VBATT_PRE	$V_{BATT}$ rising, REG04h, bit[1] = 1, $V_{BATT\_PRE}$ = 3.0V	2.9	3.0	3.1	V
Pre-charge to fast charge threshold hysteresis				90		mV
Battery auto-recharge	V <sub>RECH</sub>	Below V <sub>BATT_REG</sub> , REG04h, bit[0] = 0	60	100	140	mV
voltage threshold	V RECH	Below V <sub>BATT_REG</sub> , REG04h, bit[0] = 1	160	200	240	1110
Battery auto-recharge deglitch time	trech_dgl			200		ms
Thermal Protection	T	<b>,</b>		r	1	1
Thermal shutdown threshold <sup>(5)</sup>	T <sub>J_SHDN</sub>			150		°C
Thermal shutdown hysteresis (5)				20		°C
NTC output current	I <sub>NTC</sub>	CEB = 0, NTC = 3V	-1	0	1	μA
NTC cold temp rising threshold	V <sub>COLD</sub>	As a percentage of V <sub>DD</sub>	63	65	67	%
NTC cold temp rising threshold hysteresis				60		mV

## **ELECTRICAL CHARACTERISTICS** (continued)

 $V_{IN}$  = 5V,  $T_A$  = 25°C, unless otherwise noted.

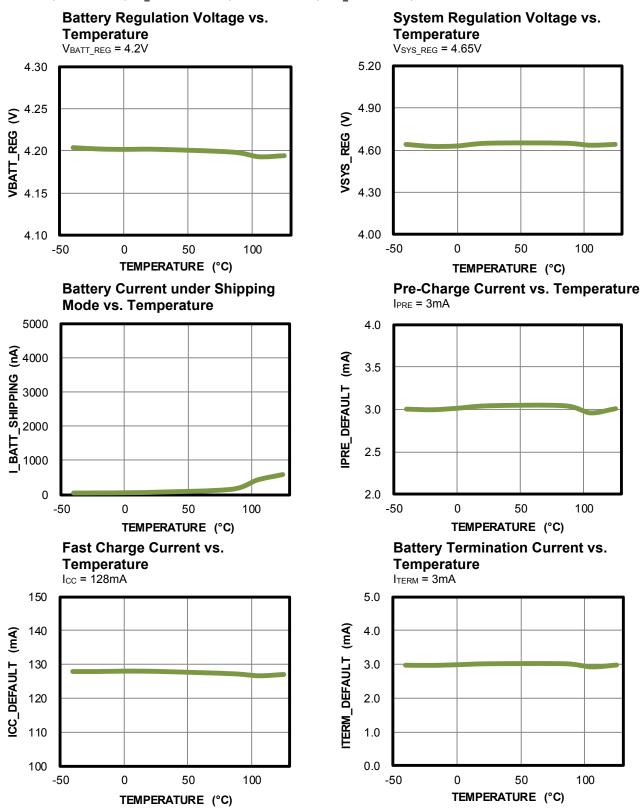
Parameter	Symbol	Condition	Min	Тур	Max	Units	
NTC hot temp falling threshold	V <sub>HOT</sub>	As a percentage of V <sub>DD</sub>	31	33	35	%	
NTC hot temp falling threshold hysteresis				70		mV	
NTC hot temp falling threshold for PCB_OTP	V <sub>HOT_PCB</sub>	As a percentage of V <sub>DD</sub>	30	32	35	%	
NTC hot temp falling threshold hysteresis for PCB_OTP				90		mV	
Logic I/O Pin Characteristic	s						
Low logic voltage threshold	$V_L$				0.4	V	
High logic voltage threshold	V <sub>H</sub>		1.3			V	
I <sup>2</sup> C Interface (SDA, SCL)							
Input high threshold level	VIH	V <sub>PULL_UP</sub> = 1.8V, SDA and SCL	1.3			V	
Input low threshold level	VIL	V <sub>PULL_UP</sub> = 1.8V, SDA and SCL			0.4	V	
Output low threshold level	V <sub>OL</sub>	I <sub>SINK</sub> = 5mA			0.4	V	
I <sup>2</sup> C clock frequency	Fscl				400	kHz	
<b>Clock Frequency and Watc</b>	Clock Frequency and Watchdog Timer						
Clock frequency	Fclk			131		kHz	
Watchdog timer	twdt	REG05h, bits[6:5] = 11		160		S	

#### NOTE:

5) Guaranteed by design

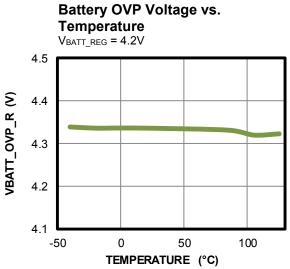
#### TYPICAL PERFORMANCE CHARACTERISTICS

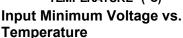
 $V_{IN}$  = 5V,  $T_A$  = 25°C,  $I_{IN\ LIM}$  = 500mA,  $I_{CC}$  = 128mA,  $V_{IN\ MIN}$  = 4.6V, unless otherwise noted.

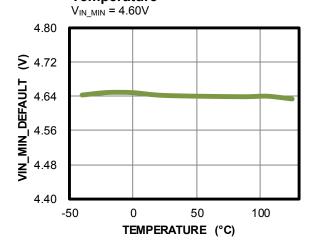


## **TYPICAL PERFORMANCE CHARACTERISTICS** (continued)

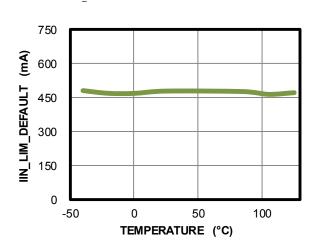
 $V_{IN}$  = 5V,  $T_A$  = 25°C,  $I_{IN\_LIM}$  = 500mA,  $I_{CC}$  = 128mA,  $V_{IN\_MIN}$  = 4.6V, unless otherwise noted.





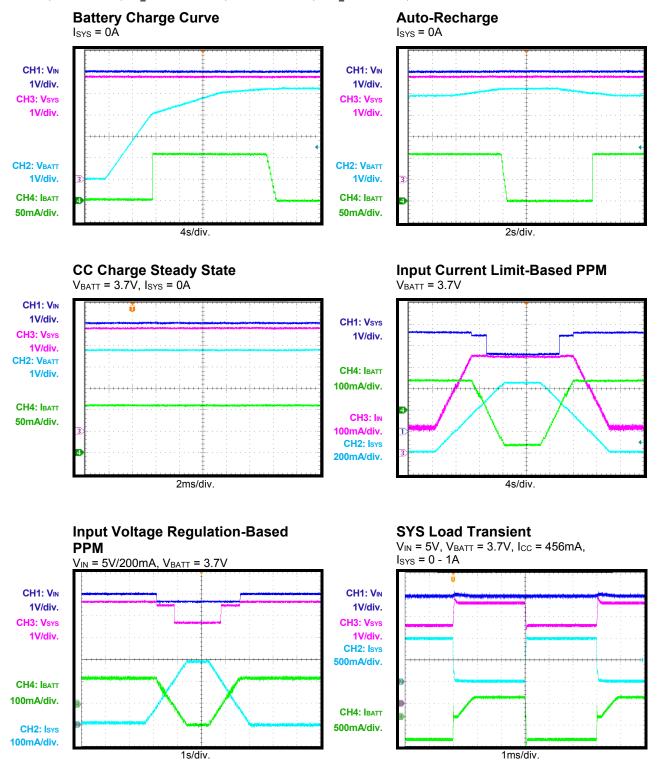


## Input Current Limit vs. Temperature $I_{IN\_LIM} = 500 mA$



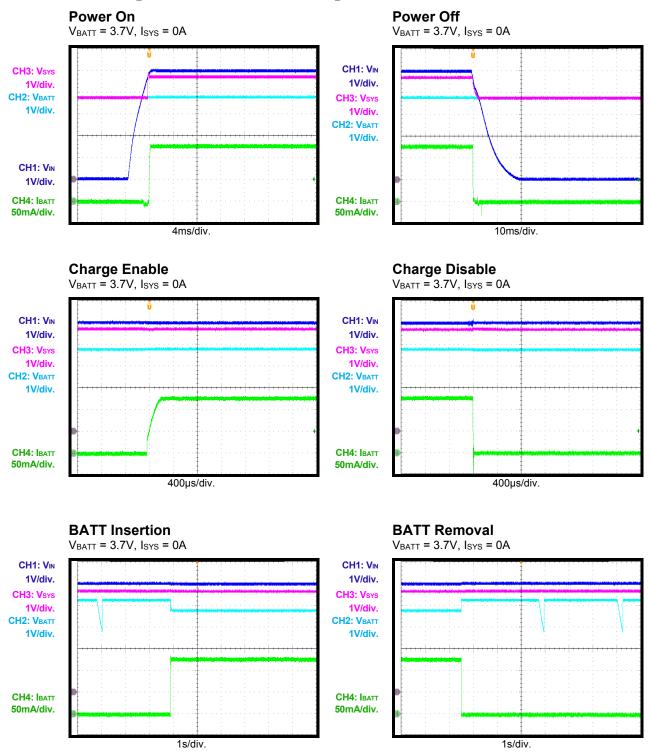
#### TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{IN}$  = 5V,  $T_A$  = 25°C,  $I_{IN\_LIM}$  = 500mA,  $I_{CC}$  = 128mA,  $V_{IN\_MIN}$  = 4.6V, unless otherwise noted.



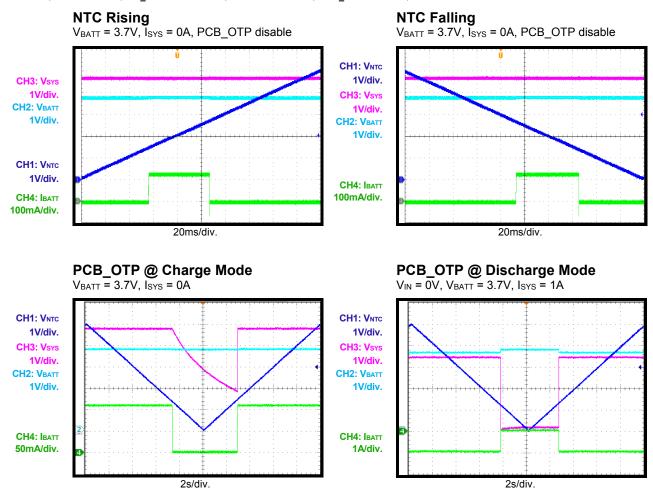
## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

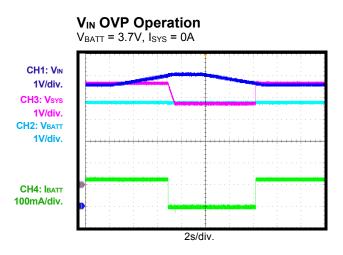
 $V_{IN}$  = 5V,  $T_A$  = 25°C,  $I_{IN\_LIM}$  = 500mA,  $I_{CC}$  = 128mA,  $V_{IN\_MIN}$  = 4.6V, unless otherwise noted.



### TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{IN}$  = 5V,  $T_A$  = 25°C,  $I_{IN\_LIM}$  = 500mA,  $I_{CC}$  = 128mA,  $V_{IN\_MIN}$  = 4.6V, unless otherwise noted.





## **FUNCTIONAL BLOCK DIAGRAM**

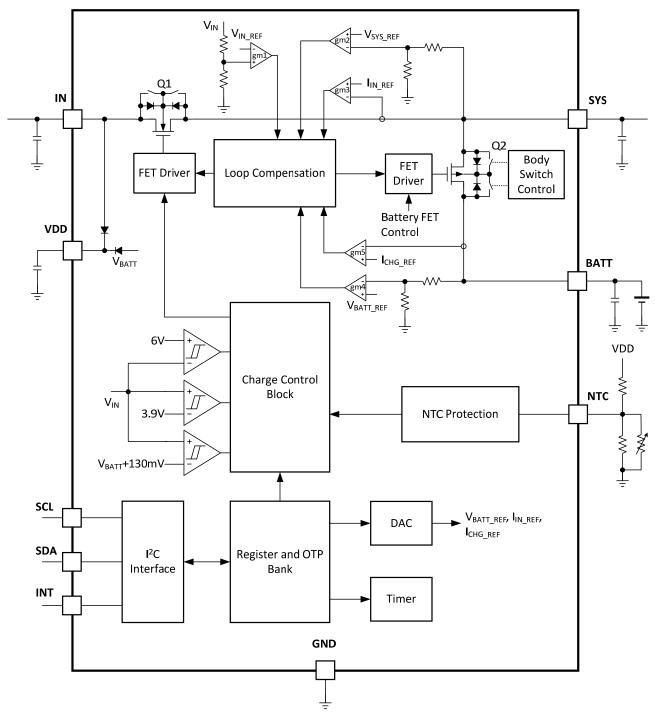


Figure 2: Functional Block Diagram

#### **OPERATION**

The MP2662 is an I<sup>2</sup>C-controlled, single-cell, Liion or Li-polymer battery charger with a complete power path management function. The full-charge function includes constant-current pre-charge, constant-current fast charge (CC) and constant voltage (CV) regulation, charge termination, auto-recharge, and a built-in timer. The power path function allows the input source to power the system and charge the battery simultaneously. The system load requirement always has priority to the charge current. When the input power is limited due to an input current limit or input voltage limit, the IC reduces the charge current automatically until the battery supplements the system load.

The IC integrates a  $290m\Omega$  LDO FET between IN and SYS and a  $100m\Omega$  battery FET between SYS and BATT.

In charging mode, the on-chip  $100m\Omega$  battery FET works as a fully featured linear charger with pre-charge, fast charge, constant voltage charge, charge termination, auto-recharge, NTC monitoring, built-in timer control, and thermal protection. The charge current can be programmed via the  $I^2C$  interface. The IC adjusts the charge current when the die temperature exceeds the thermal regulation threshold ( $120^{\circ}C$  default).

In supplement mode, the  $100m\Omega$  battery FET is turned on to connect the battery to the system load when the input power is not sufficient enough to power the system load. When the input is removed, the  $100m\Omega$  battery FET is also fully turned on to allow the battery to power up the system.

The system load is satisfied in priority, and the remaining current is used to charge the smart power path management battery. The MP2662 reduces the charging current or uses power from the battery to satisfy the system load when its demand is over the input power capacity.

Figure 3 shows the power path management structure for the MP2662.

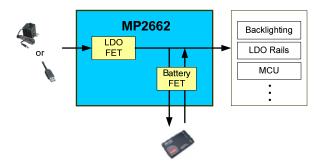


Figure 3: Power Path Management Structure

#### **Power Supply**

The internal bias circuit of the IC is powered from the higher voltage of either IN or BATT. When IN or BATT rises above its respective under-voltage lockout (UVLO) threshold, the sleep comparator, battery depletion comparator, and battery FET driver are all active. The I<sup>2</sup>C interface is ready for communication, and all registers are reset to the default value. The host can access all of the registers.

#### Input OVP and UVLO

The MP2662 has an input over-voltage protection (OVP) threshold and input UVLO threshold. Once the input voltage is out of its normal range, the LDO FET (Q1) is turned off immediately.

When the input voltage is identified as a good source, a 200µs immunity timer becomes active. If the input power is normal until the 200µs expires, the system starts up. Otherwise, Q1 remains off (see Figure 4).

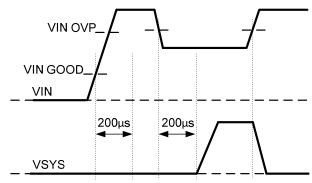


Figure 4: Input Power Detection Operation
Profile

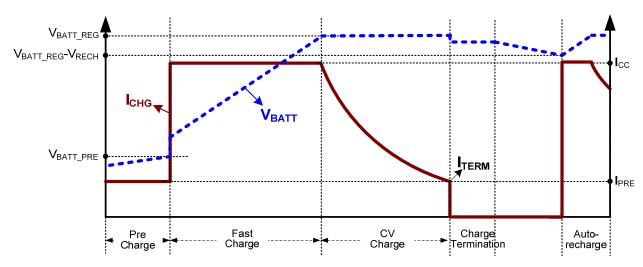


Figure 5: Battery Charge Profile

#### **Power Path Management**

The MP2662 employs a pass-through power path structure with the battery FET (Q2) to decouple the system from the battery. This allows for separate control between the system and the battery. The system is given the priority to start up, even with a deeply discharged or missing battery. When the input power is available, even with a depleted battery, the system voltage is always regulated to  $V_{\text{SYS\_REG}}$  by the integrated LDO FET.

The direct power structure consists of a frontend LDO FET between IN and SYS and a battery FET between SYS and BATT. The LDO FET and battery FET can be controlled by the I<sup>2</sup>C.

Table 1: FET Control via I<sup>2</sup>C

FET On/Off Changed by	Hi-Z Mode and Charge Cont				
Control	Set EN_HIZ to 1	Set CEB to 1			
LDO FET	OFF	х			
Battery FET (charging)	х	OFF			
Battery FET (discharging)	х	х			

**NOTE:** x = Don't care.

For the system voltage control, when the input voltage is higher than  $V_{\text{SYS\_REG}}$ , the system voltage is regulated to  $V_{\text{SYS\_REG}}$ . When the input voltage is lower than  $V_{\text{SYS\_REG}}$ , the LDO FET is fully on with the input current limit.

 $V_{\text{SYS\_REG}}$  can be programmed through REG07h, bits[3:0].

#### **Battery Charge Profile**

The MP2662 provides three main charging phases: pre-charge, fast-current charge, and constant-voltage charge (see Figure 5).

- 1. Phase 1 (pre-charge): The MP2662 can safely pre-charge the deeply depleted battery until the battery voltage reaches the pre-charge to fast-charge threshold (VBATT\_PRE). The pre-charge current is also programmed through REG03h, bits[3:0]. If VBATT\_PRE is not reached before the pre-charge timer (1hr) expires, the charge cycle stops, and a corresponding timeout fault signal is asserted.
- 2. Phase 2 (fast charge): When the battery voltage exceeds V<sub>BATT\_PRE</sub>, the MP2662 enters a fast-charge phase. The fast-charge current can be programmed via REG02h, bits[5:0].
- 3. Phase 3 (constant-voltage charge): When the battery voltage rises to the battery-full voltage (V<sub>BATT\_REG</sub>) set via REG04h, bits[7:2], the charge mode changes from CC mode to CV mode, and the charge current starts decreasing.

Assuming that the termination function EN\_TERM is set via REG05h, bit[4] = 1, the charge cycle is considered to be completed when the charge current ( $I_{CHG}$ ) reaches the termination current threshold ( $I_{TERM}$ ) and a 3.2s delay timer is initiated. During this 3.2s delay period,  $I_{CHG}$  is always smaller than  $I_{TERM}$  +  $I_{TERM\_HYS}$ .

The charge status is updated to charge done once the 3.2s delay timer expires.

The termination charge current threshold (I<sub>TERM</sub>) can be programmed via REG03h, bits[3:0].

The charge current can also be terminated when the termination conditions are met if TERM\_TMR set via REG05h, bit[0] = 0. Otherwise, the charge current continues to taper off.

If EN\_TERM = 0, the termination function is disabled and all of the above actions are invalid (see Table 2).

**Table 2: Termination Function Selection Table** 

EN_	TERM_	After IBATT ITERM IN C	
TERM	TMR	Operation	Charge Status
0	х	Keep CV charge	Charge
1	0	Charge done	Charge done
1	1	Keep CV charge	Charge

**NOTE:** x = Don't care.

During the entire charging process, the actual charge current may be less than the register setting due to other loop regulations, such as dynamic power management (DPM) regulation (input voltage, input current) or thermal regulation.

A new charge cycle starts when any of the following conditions are valid:

- The input power is recycled.
- Battery charging is enabled via the I<sup>2</sup>C.
- Auto-recharge kicks in.

Under the following conditions:

- No thermistor fault at NTC.
- · No safety timer fault.

- No battery over-voltage event.
- Battery FET is not forced off.

#### **Automatic Recharge**

When the battery is fully charged and charging is terminated, the battery may be discharged due to system consumption or self-discharge. When the battery voltage is discharged below the recharge threshold and  $V_{\text{IN}}$  is still in the operating range, the MP2662 begins another new charging cycle automatically without having to restart a charging cycle manually.

The auto-recharge function is valid only when EN TERM = 1 and TERM TMR = 0.

#### **Battery OVP**

The MP2662 is designed with a built-in battery over-voltage limit (about 130mV higher than  $V_{BATT\_REG}$ ). When a battery over-voltage event occurs, the MP2662 suspends charging immediately and asserts a fault.

## Input Current- and Input Voltage-Based Power Management

To meet the input source's (typically USB) maximum current limit specification, the MP2662 uses an input current-based power management by monitoring the input current continuously. The total input current limit can be programmed via the I<sup>2</sup>C to prevent the input source from being over-loaded.

If the preset input current limit is higher than the rating of the input source, the back-up input voltage-based power management also works to prevent the input source from being overloaded. Either the input current limit or the input voltage limit is reached, the Q1 FET between IN and SYS are regulated so that the total input power is limited. As a result, the system voltage drops. Once the system declines to a minimum value of  $V_{\text{SYS\_REG}}$  - 135mV and  $V_{\text{IN}}$  - 175mV, the charge current is reduced to prevent the system voltage from dropping further.

The voltage-based DPM regulates the input voltage to  $V_{\text{IN\_MIN}}$  when the load is over the input power capacity.

 $V_{\text{IN\_MIN}}$  set via the I<sup>2</sup>C should be at least 250mV higher than  $V_{\text{BATT\_REG}}$  to ensure the stable operation of the regulator. The input voltage limit function can be disabled by REG07h, bit[6].

#### **Battery Supplement Mode**

The charge current is reduced to keep the input current or input voltage in regulation when DPM occurs. If the charge current is reduced to zero and the input source is still overloaded due to a heavy system load, the system voltage begins decreasing. Once the system voltage drops to 37mV below the battery voltage, the MP2662 enters battery supplement mode, and the ideal diode mode is enabled. The battery FET is regulated to keep  $V_{\text{BATT}}$  -  $V_{\text{SYS}}$  at 30mV when I<sub>DSCHG</sub> (supplement current) \* R<sub>ON BATT</sub> is lower than 30mV. In the case that I<sub>DSCHG</sub> \* R<sub>ON BATT</sub> is higher than 30mV, the battery FET is fully turned on to maintain the ideal forward voltage. When the system load decreases, once  $V_{SYS}$  is higher than V<sub>BATT</sub> + 20mV, the ideal diode mode is disabled. Figure 6 shows the dynamic power management and battery supplement mode operation profile.

When  $V_{\text{IN}}$  is not available, the MP2662 operates in discharge mode, and the battery FET is always fully on to reduce loss.

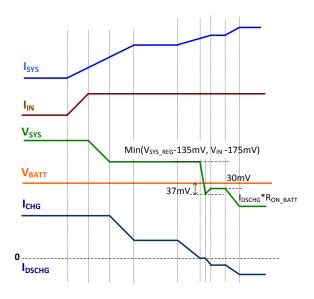


Figure 6: Dynamic Power Management and Battery Supplement Operation Profile

#### **Battery Regulation Voltage**

The battery voltage for the constant voltage regulation phase is  $V_{BATT\_REG}$ . When  $V_{BATT\_REG}$  = 4.2V, it has a ±0.5% accuracy over the ambient temperature range of 0°C to 50°C.

#### Thermal Regulation and Thermal Shutdown

The MP2662 monitors the internal junction temperature continuously to maximize power delivery and avoid overheating the chip. When the internal junction temperature reaches the preset limit of  $T_{J\_REG}$  (120°C default), the IC starts to reduce the charge current to prevent higher power dissipation. The multiple thermal regulation thresholds from 60°C to 120°C help the system design meet the thermal requirements in different applications. The junction temperature regulation threshold can be set via REG07h, bits[5:4].

When the junction temperature reaches 150°C, both Q1 and Q2 turn off.

#### Negative Temperature Coefficient (NTC) Temperature Sensor

NTC allows the MP2662 to sense the battery temperature using the thermistor (usually available in the battery pack) to ensure a safe operating environment for the battery. Connect appropriately valued resistors from VDD to NTC to ground. The resistor divider works with a thermistor connected from NTC to ground. The NTC voltage is determined by the resistor divider whose divide ratio depends on the temperature. The M2662 sets a pre-determined upper and lower bound of the divide ratio for NTC cold and NTC hot internally.

In the MP2662, the  $I^2C$  default setting is PCB\_OTP. This function can be changed through the  $I^2C$  (see Table 3).

 I²C Control

 EN\_NTC
 EN\_PCB\_OTP

 0
 x
 Disable

 1
 1
 NTC

 1
 0
 PCB\_OTP

**Table 3: NTC Function Selection Table** 

**NOTE:** x = Don't care.

When PCB\_OTP is selected and the NTC voltage is lower than the NTC hot threshold, both the LDO FET and battery FET are off. The PCB\_OTP fault also sets the NTC\_FAULT status (REG09h, bit[1]) to 1 to show the fault. Operation resumes once the NTC voltage is higher than the NTC hot threshold.

The NTC function works only in charge mode. Once the NTC voltage falls out of this divide ratio, the temperature is outside of the safe operating range, and the MP2662 stops charging and report it on the status bits. Charging resumes automatically after the temperature falls back into the safe range.

#### **Safety Timer**

The MP2662 provides both a pre-charge and fast-charge safety timer to prevent an extended charging cycle due to abnormal battery conditions. The safety timer is 1hr when the battery voltage is lower than  $V_{\text{BATT\_PRE}}$ . The fast-charge safety timer starts when the battery enters fast-charge mode. The fast-charge safety timer can be programmed through the  $I^2C$ . The safety timer can be disabled via the  $I^2C$ .

The following actions can restart the safety timer:

- A new charge cycle is initiated
- Charge enable toggling
- HIZ disable toggling

#### **Host Mode and Default Mode**

The MP2662 is a host-controlled device. After the power-on reset, the MP2662 starts up in a watchdog timer expiration state or default mode. All registers are in their default settings.

The watchdog timer works in both charge and discharge mode. When the watchdog timer out, most registers return to the default value (refer to the  $I^2C$  Register Map). When the watchdog timer is out in both charge and discharge mode, both the LDO FET and battery FET are turned off. They turn on again automatically after  $t_{RST\_DUR}$ , which can be programmed by REG01h, bit[5].

To save quiescent current during discharge mode, the watchdog timer can be turned off during by setting REG05h, bit[7] to 0.

Any write to the MP2662 switches it to host mode. All charge parameters are

programmable. If the watchdog timer (REG05h, bits[6:5]) is not disabled, the host must reset the watchdog timer regularly by writing 1 to REG02h, bit[6] before the watchdog timer expires to keep the device in host mode. Once the watchdog timer expires, the MP2662 goes back to default mode. The watchdog timer limit can also be programmed or disabled by the host control.

When the REG05h, bits[6:5] is set to 00, then the watchdog timer is disabled under both charge mode and discharge mode regardless of the status of REG05h, bit[7].

Operation can also be switched to default mode when one of the following conditions are valid:

- Refresh input without battery.
- Re-insert battery with no V<sub>IN</sub>.
- Register REG02h, bit[7] is reset.

#### **Battery Discharge Function**

If the battery is connected and the input source is missing, the battery FET is fully on when  $V_{BATT}$  is above the  $V_{BATT\_UVLO}$  threshold. The  $100m\Omega$  battery FET minimizes conduction loss during discharge. The quiescent current of the MP2662 is as low as  $6.5\mu A$  in this mode. The low on resistance and low quiescent current help extend the running time of the battery.

#### **Over-Discharge Current Protection**

The MP2662 has an over-discharge current protection in discharge mode and supplement mode. Once I<sub>DSCHG</sub> exceeds the programmable discharge current limit (2A default), the battery FET turns off after a 60µs delay. The MP2662 enters hiccup mode as part of the over-current protection (OCP). The discharge current limit can be programmed high to 3.2A through the I<sup>2</sup>C. If the discharge current goes high and reaches the internal fixed current limit (about 3.7A), the battery FET turns off and begins hiccup mode immediately.

Similarly, when the battery voltage falls below the programmable  $V_{BATT\_UVLO}$  threshold (2.76V default), the battery FET turns off to prevent an over-discharge.

#### **System Short-Circuit Protection (SCP)**

The MP2662 features SYS node short-circuit protection (SCP) for both the IN to SYS path and the BATT to SYS path.

The system voltage is monitored continuously. If  $V_{SYS}$  is lower than 1.5V, system SCP for both the IN to SYS path and the BATT to SYS path is active.  $I_{DSCHG}$  decreases to half of the original value.

For the IN to SYS path, once  $I_{\rm IN}$  is over the protection threshold, both the LDO FET and battery FET are turned off immediately, and the MP2662 enters hiccup mode. Otherwise, the maximum current limit is not reached. When  $V_{\rm SYS}$  is lower than 1.5V and the setting input current limit is reached, the hiccup mode also starts after a 60µs delay. The hiccup mode interval is 800µs.

For the BATT to SYS path, once  $I_{BATT}$  is over the 3.7A protection threshold, both the LDO FET and battery FET are turned off immediately, and the MP2662 enters hiccup mode. When the battery discharge current limit threshold is reached, hiccup mode starts after a 60µs delay. The hiccup mode interval is 800µs.

For details, please refer to the flow chart shown in Figure 20.

Particularly, if a system short-circuit occurs when both the input and battery are present, the protection mechanism of both paths works. The faster one of the two dominates the hiccup operation.

#### **Interrupt to Host (INT)**

The MP2662 also has an alert mechanism that can output an interrupt signal via INT to notify the system of the operation by outputting a 256µs low-state INT pulse. All of the below events can trigger an INT output:

- Good input source detected (PG\_STAT)
- Charge completed
- Charging status change
- Faults in REG09h (watchdog timer fault, thermal fault, safety timer fault, battery OVP fault, NTC fault)

When fault occurs, the MP2662 sends out an INT pulse and latches the fault state in REG09h. After the MP2662 exits the fault state, the fault bit is reset to 0 after the host reads REG09h. The NTC fault bit is not latched and always reports the current thermistor conditions.

The INT signal can be masked when the corresponding control bit is set in REG06h, bits[4:0]. When an INT condition is masked, this means that the INT pin signal (and register bit) will not trigger when the corresponding condition occurs. Masking INTs is useful when writing software code to avoid unnecessary interruptions due to these events.

#### **Battery Disconnection Function**

In applications where the battery is not removable, it is essential to disconnect the battery from the system for shipping mode or to allow the system power to be reset during the application. The MP2662 provides both shipping mode (shown in Table 4) and system reset mode for different applications.

**Table 4: Shipping Mode Control** 

lt a man		Enter Shipping Mode	Exit Shipp	oing Mode
	Items	Set FET_DIS to 1	o 1 INT H to L for 2s V <sub>IN</sub> Plug	
	LDO FET	Х	х	On
Batte	ry FET (charging)	Off (tsmen_dgl later)	On	On (2s later)
Battery	/ FET (discharging)	Off (t <sub>SMEN_DGL</sub> later)	On	On (2s later)

**NOTE:** x = Don't care.

The IC has a register bit for battery disconnection control (FET\_DIS). If this bit is set to 1, the MP2662 enters shipping mode after a delay time, which can be programmed

by REG09h, bits[7:6]. The battery FET turns off, and the FET\_DIS bit refreshes to 0 after the battery FET turns off. Pull the INT pin down or

plug in the input adapter for 2s to wake the MP2662 up from shipping mode.

The MP2662 can also reuse an INT pin to cut off the path from the battery to the system under the condition needed to reset the system manually. Once the logic at INT is set low for longer than t<sub>RST DGL</sub> (which can be programmed REG01h. bits[7:6]), the batterv is disconnected from the system by turning off the battery FET. The off state lasts for t<sub>RST DUR</sub>, which can be programmed by REG01h, bit[5]. Then the battery FET is turned on automatically, and the system is powered by the battery again. During the off period, the INT pin is not limited to be high or low.

The MP2662 can reset the system by controlling the INT pin (see Figure 7).

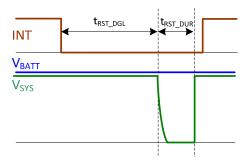


Figure 7: System Reset Function Operation
Profile

#### **SERIAL INTERFACE**

The IC uses an I<sup>2</sup>C-compatible interface to set the charging parameters and instantaneously report the device status. The I<sup>2</sup>C is a two-wire serial interface. Only two bus lines are required: a serial data line (SDA) and serial clock line (SCL). Both the SDA and SCL lines are open drains that must be connected to the positive supply voltage via a pull-up resistor.

The IC operates as a slave device and receives control inputs from the master device, such as a microcontroller. The SCL line is always driven by the master device. The I<sup>2</sup>C interface supports both standard mode (up to 100kbits/s) and fast mode (up to 400kbits/s).

All transactions begin with a start (S) condition and are terminated by a stop (P) condition. Start and stop conditions are always generated by the master. A start condition is defined as a high-to-low transition on the SDA line while SCL

is high. A stop condition is defined as a low-tohigh transition on the SDA line while the SCL is high (see Figure 8).

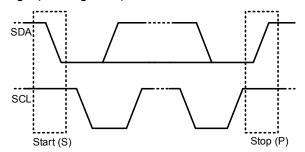


Figure 8: Start and Stop Conditions

For data validity, the data on the SDA line must be stable during the high period of the clock. The high or low state of the SDA line can only change when the clock signal on the SCL line is low (see Figure 9). Every byte on the SDA line must be 8 bits long. The number of bytes that can be transmitted per transfer is unrestricted. Data is first transferred with the most significant bit (MSB).

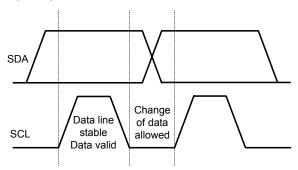


Figure 9: Bit Transfer on the I2C Bus

Each byte has to be followed by an Acknowledge (ACK) bit which is generated by the receiver, to signal the transmitter that the byte was successfully received.

The ACK signal occurs when the transmitter releases the SDA line during the acknowledge clock pulse. This allows the receiver to pull the SDA line low. The SDA line stays low during the high period of the ninth clock.

If the SDA line is high during the ninth clock, this is defined as a not acknowledge (NACK) signal. The master can then generate either a stop condition to abort the transfer or a repeated start condition to start a new transfer.

After the start condition, a slave address is sent. This address is 7 bits long, followed by the eighth data direction bit (R/W). A 0 indicates a transmission (write), and a 1 indicates a request for data (read). Figure 10 shows the address bit arrangement.

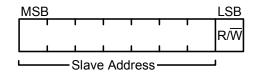


Figure 10: 7-Bit Address

Figures 11–15 show detailed signal sequences.

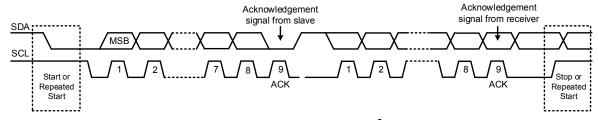


Figure 11: Data Transfer on the I<sup>2</sup>C Bus

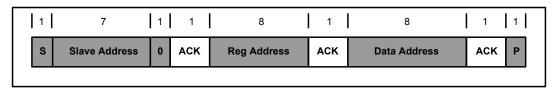


Figure 12: Single Write



Figure 13: Single Read

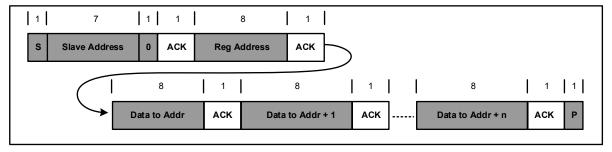
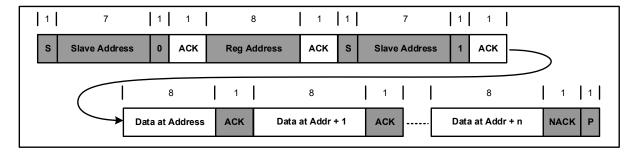


Figure 14: Multi-Write



## I<sup>2</sup>C REGISTER MAP

IC Address: REG07h (reserved some trim options)

Register Name	Address	R/W	Description	Default
REG00h	0x00	r/w	Input source control register	1001 1111
REG01h	0x01	r/w	Power on configuration register	1010 1100
REG02h	0x02	r/w	Charge current control register	0000 1111
REG03h	0x03	r/w	Discharge/termination current	1001 0001
REG04h	0x04	r/w	Charge voltage control register	1010 0011
REG05h	0x05	r/w	Charge termination/timer control register	0011 1010
REG06h	0x06	r/w	Miscellaneous operation control register	1100 0000
REG07h	0x07	r/w	System voltage regulation register	0011 1001
REG08h	0x08	r	System status register	0100 0000
REG09h	0x09	r/w	Fault register	0000 0000
REG0Ah	0x0A	N/A	Address register	1110 0000

**REG 00h (Default: 1001 1111)** 

	2000. (2000.00.00.)										
Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment				
7	V <sub>IN_MIN</sub> [3]	1	Υ	N	r/w	640mV					
6	V <sub>IN_MIN</sub> [2]	0	Υ	Ν	r/w	320mV	Offset: 3.88V				
5	V <sub>IN_MIN</sub> [1]	0	Υ	Ν	r/w	160mV	Range: 3.88V - 5.08V Default: 4.60V (1001)				
4	V <sub>IN_MIN</sub> [0]	1	Υ	Ν	r/w	80mV					
3	I <sub>IN_LIM</sub> [3]	1	Υ	Ν	r/w	240mA					
2	I <sub>IN_LIM</sub> [2]	1	Υ	N	r/w	120mA	Offset: 50mA				
1	I <sub>IN_LIM</sub> [1]	1	Υ	Ν	r/w	60mA	Range: 50mA - 500mA Default: 500mA (1111)				
0	I <sub>IN_LIM</sub> [0]	1	Υ	N	r/w	30mA					

## **REG 01h (Default: 1010 1100)**

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	t <sub>RST_DGL</sub> [1]	1	Υ	Υ	r/w	00: 8s 01: 12s	Pull INT low to disconnect the battery.
6	t <sub>RST_DGL</sub> [0]	0	Υ	Υ	r/w	10: 16s 11: 20s	Default: 16s (10)
5	trst_dur	1	Y	Y	r/w	0: 2s 1: 4s	Battery FET is off for a period of time before autoon  Default: 4s (1)
4	EN_HIZ <sup>(6)</sup>	0	Υ	Υ	r/w	0: Disable 1: Enable	Default: Disable (0)
3	CEB	1	Y	Υ	r/w	0: Charge enable 1: Charge disable	Charge configuration Default: Charge disable (1)
2	V <sub>BATT_UVLO</sub> [2]	1	Υ	Υ	r/w	360mV	Battery UVLO threshold
1	V <sub>BATT_UVLO</sub> [1]	0	Υ	Υ	r/w	180mV	Offset: 2.4V Range: 2.4V - 3.03V
0	VBATT_UVLO [0]	0	Υ	Υ	r/w	90mV	Default: 2.76V (100)

#### NOTE:

## REG 02h (Default: 0000 1111)

	· · · · · · · · · · · · · · · · · · ·		, ,			T	· · · · · · · · · · · · · · · · · · ·
Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	Register Reset	0	Y	N	r/w	0: Keep current setting 1: Reset	Default: Keep current register setting (0)
6	I <sup>2</sup> C Watchdog Timer Reset	0	Υ	Υ	r/w	0: Normal 1: Reset	Default: Normal (0)
5	Icc [5]	0	Υ	Υ	r/w	256mA	
4	Icc [4]	0	Υ	Υ	r/w	128mA	Fast charge current setting.
3	I <sub>CC</sub> [3]	1	Υ	Υ	r/w	64mA	Offset: 8mA
2	I <sub>CC</sub> [2]	1	Υ	Υ	r/w	32mA	Range: 8mA (000000) - 456mA (111000)
1	Icc [1]	1	Υ	Υ	r/w	16mA	Default: 128mA (001111)
0	Icc [0]	1	Υ	Υ	r/w	8mA	

<sup>6)</sup> This bit only controls the on and off function of the LDO FET.

## **REG 03h (Default: 1001 0001)**

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	IDSCHG [3]	1	Υ	Υ	r/w	1600mA	BATT to SYS discharge current limit.
6	I <sub>DSCHG</sub> [2]	0	Υ	Υ	r/w	800mA	Offset: 200mA
5	I <sub>DSCHG</sub> [1]	0	Υ	Υ	r/w	400mA	Range: 400mA - 3.2A
4	IDSCHG [0]	1	Υ	Υ	r/w	200mA	Valid range: 0001 - 1111 Default: 2000mA (1001)
3	I <sub>TERM</sub> [3]	0	Υ	Υ	r/w	16mA	Termination current.
2	I <sub>TERM</sub> [2]	0	Υ	Υ	r/w	8mA	Offset: 1mA
1	I <sub>TERM</sub> [1]	0	Υ	Υ	r/w	4mA	Range: 1mA - 31mA
0	I <sub>TERM</sub> [0]	1	Υ	Υ	r/w	2mA	Default: 3mA (0001)

## **REG 04h (DEFAULT: 1010 0011)**

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	V <sub>BATT_REG</sub> [5]	1	Υ	Υ	r/w	480mV	
6	V <sub>BATT_REG</sub> [4]	0	Υ	Υ	r/w	240mV	Battery regulation voltage.
5	VBATT_REG [3]	1	Υ	Υ	r/w	120mV	Offset: 3.60V
4	V <sub>BATT_REG</sub> [2]	0	Υ	Υ	r/w	60mV	Range: 3.60V - 4.545V
3	V <sub>BATT_REG</sub> [1]	0	Υ	Υ	r/w	30mV	Default: 4.2V (101000)
2	VBATT_REG [0]	0	Υ	Υ	r/w	15mV	
1	V <sub>BATT_PRE</sub>	1	Υ	Υ	r/w	0: 2.8V 1: 3.0V	Pre-charge to fast charge threshold.
						1. 3.0 V	Default: 3.0V (1)
0	VRECH	1	Y	Υ	r/w	0: 100mV 1: 200mV	Battery recharge threshold (below V <sub>BATT_REG</sub> ).
						1. 2001110	Default: 200mV (1)

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment	
7	EN_WD_DISCHG	0	Υ	Ν	r/w	0: Disable 1: Enable	Watchdog control in discharge mode.	
							Default: Disable (0)	
6	WATCHDOC [4]	0	Y	N	r/w		I <sup>2</sup> C watchdog timer limit.	
0	WATCHDOG [1]	U	Ť	IN	1/W	00: Disable timer	Default: 40s (01)	
5	WATCHDOG [0]	1	Y	N	r/w	01: 40s 10: 80s 11: 160s	If Bit[6:5] = 00, then watchdog timer is disabled regardless of whether bit[7] is set or not.	
4	EN_TERM	1	Y	Υ	r/w	0: Disable 1: Enable	Termination setting (controlling the termination is allowed or not).	
							Default: Enable (1)	
3	EN TIMED	1	Y	Υ	r/w	0: Disable	Safety timer setting.	
3	EN_TIMER	ı	Ť	Ť	1/W	1: Enable	Default: Enable timer (1)	
2	CHG_TMR [1]	0	Υ	Υ	r/w	00: 3hrs	Fast charge timer.	
1	CHG_TMR [0]	1	Y	Y	r/w	01: 5hrs 10: 8hrs 11: 12hrs	Default: 5hrs (01)	
0	TERM_TMR	0	Y	Υ	r/w	0: Disable 1: Enable	Termination timer control (when TERM_TMR is enabled, the IC will not suspend the charge current after the charge termination).  Default: (0)	

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	EN_NTC	1	Y	Y	r/w	0: Disable 1: Enable	Default: Enable (1)
6	TMR2X_EN	1	Y	Y	r/w	0: Disable 2X extended safety timer during PPM 1: Enable 2X extended safety timer during PPM	Default: Enable (1)
5	FET_DIS (7)	0	Y	N	r/w	0: Enable 1: Turn off	Default: Enable (0)
4	PG_INT_Control	0	Υ	Y	r/w	0: On 1: Off	Default: On (0)
3	EOC_INT_Control	0	Υ	Y	r/w	0: On 1: Off	Charge completed INT mask control
							Default: On (0)
2	CHG STATUS_ INT_Control	0	Y	Υ	r/w	0: On 1: Off	Charging status change INT mask control (charging status contain: not charging, pre charge and charge).
							Default: On (0)
1	NTC_INT_Control	0	Y	Y	r/w	0: On 1: Off	Default: On (0)
0	BATTOVP_INT_Control	0	Y	Y	r/w	0: On 1: Off	Default: On (0)

#### NOTE:

<sup>7)</sup> This bit only controls the turn off function of the battery FET, including charge and discharge.

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	EN PCB OTP	0	Υ	Υ	r/w	0: Enable	PCB_OTP enable.
,		)	'		17 44	1: Disable	Default: Enable (0)
6	EN_VINLOOP	0	Υ	Υ	r/w	0: Enable 1: Disable	Default: Enable (0)
5	T <sub>J_REG</sub> [1]	1	Υ	Υ	r/w	00: 60°C	Thermal regulation
4	T <sub>J_REG</sub> [0]	1	Υ	Υ	r/w	01: 80°C 10: 100°C 11: 120°C	threshold.  Default: 120°C (11)
3	V <sub>SYS_REG</sub> [3]	1	Υ	N	r/w	400mV	System voltage
2	Vsys_reg [2]	0	Υ	N	r/w	200mV	regulation.
1	Vsys_reg [1]	0	Υ	N	r/w	100mV	Offset: 4.2V Range: 4.2V - 4.95V
0	Vsys_reg [0]	1	Υ	N	r/w	50mV	Default: 4.65V (1001)

## REG 08h (Default: 0100 0000)

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment	
7	WATCHDOG_FAULT	0	N/A	N/A	r	0: Normal 1: Watchdog timer expiration	Normal (0)	
6	Rev [1]	1	N/A	N/A	r	00: reserved	Revision number.	
5	Rev [0]	0	N/A	N/A	r	01: reserved 10: MP2662 11: reserved	Default: (10)	
4	CHG_STAT [1]	0	N/A	N/A	r	00: Not charging		
3	CHG_STAT [0]	0	N/A	N/A	r	01: Pre charge 10: Charge 11: Charge done	Not charging (00)	
2	PPM_STAT	0	N/A	N/A	r	0: No PPM 1: In PPM	No PPM (0)	
1	PG_STAT	0	N/A	N/A	r	0: Power fail 1: Power good	Power fail (0)	
0	THERM_STAT	0	N/A	N/A	r	0: No thermal regulation 1: In thermal regulation	No thermal regulation (0)	

## **REG 09h (Default: 0000 0000)**

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment	
7	EN_SHIPPING_DGL[1]	0	Υ	N	r/w	00: 1s 01: 2s	Enter shipping mode deglitch time	
6	EN_SHIPPING_DGL[0]	0	Υ	N	r/w	10: 4s 11: 8s	Default: 1s (00)	
5	VIN_FAULT	0	N/A	N/A	r	0: Normal 1: Input fault (OVP or bad source)	Normal (0)	
4	THEM_SD	0	N/A	N/A	r	0: Normal 1: Thermal shutdown	Normal (0)	
3	BAT_FAULT	0	N/A	N/A	r	0: Normal 1: Battery OVP	Normal (0)	
2	STMR_FAULT	0	N/A	N/A	r	0: Normal 1: Safety timer expiration	Normal (0)	
1	NTC_FAULT [1]	0	N/A	N/A	r	0: Normal 1: NTC hot	Normal (0)	
0	NTC_FAULT [0]	0	N/A	N/A	r	0: Normal 1: NTC cold	Normal (0)	

## REG 0Ah (Default: 1110 0000) (8)

Bit	Name	POR	Reset by REG_RST	Reset by WTD	R/W	Description	Comment
7	ADDR[2]	1	N/A	N/A	N/A	001: 01h 010: 02h	
6	ADDR[1]	1	N/A	N/A	N/A	011: 03h   100: 04h   101: 05h	IC Address. Default: 111 (07h)
5	ADDR[0]	1	N/A	N/A	N/A	110: 06h 111: 07h	
4	Reserved	0	N/A	N/A	N/A		
3	Reserved	0	N/A	N/A	N/A		
2	Reserved	0	N/A	N/A	N/A		
1	Reserved	0	N/A	N/A	N/A		
0	Reserved	0	N/A	N/A	N/A		

#### NOTE:

<sup>8)</sup> This register is for one-time programming only and is not accessible.

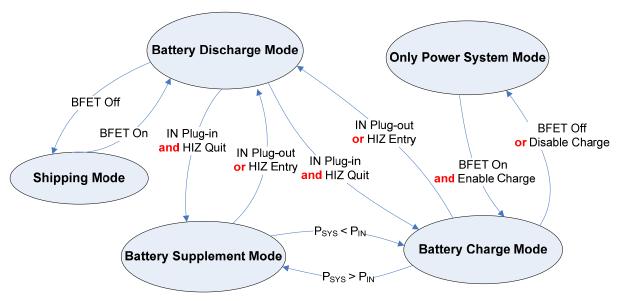
## **ONE-TIME PROGRAMMING MAP**

#	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
0x01		N/A			CEB		N/A		
0x02		N/A		Icc: 8mA - 456mA / 8mA step					
0x03		N/A		I <sub>TERM</sub> : 1mA - 31mA/2mA step					
0x04		V <sub>BATT_REG</sub> : 3	3.6V - 4.545	5V / 15mV st	tep		N	/A	
0x05	N/A	WATCHDO	G	N/A					
0x07	N/A	EN_VINLOOP			N/A	Ą			
0x0A		Address	•			N/A	•		

## **ONE-TIME PROGRAMMING DEFAULT**

One-Time Programmable Items	Default
CEB	Charge disable
Icc	128mA
ITERM	3mA
V <sub>BATT_REG</sub>	4.2V
WATCHDOG	40s
EN_VINLOOP	Enable
Address	07h

### STATE CONVERSION CHART



**Figure 16: State Machine Conversion** 

### **CONTROL FLOWCHART**

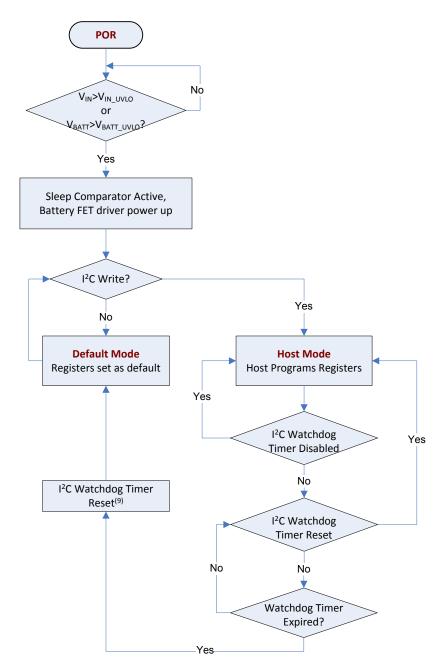


Figure 17: Default Mode and Host Mode Selection

#### NOTE:

9) Once the watchdog timer expires, the I<sup>2</sup>C watchdog timer must be reset or will not be valid in the next cycle.

## **CONTROL FLOWCHART** (continued)

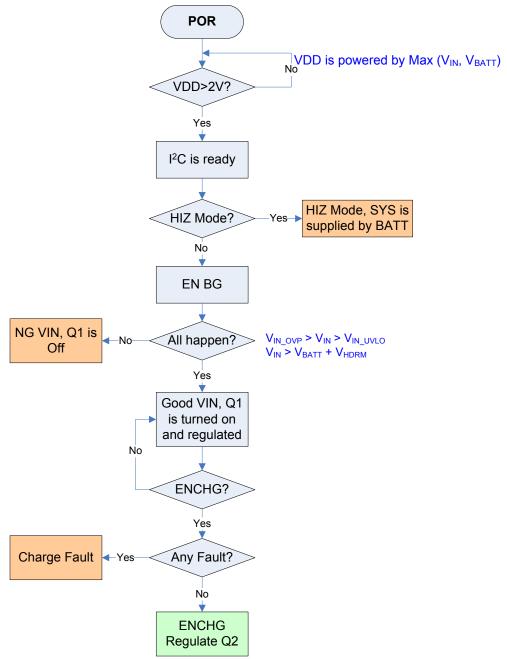


Figure 18: Input Power Start-Up Flow Chart

## **CONTROL FLOWCHART** (continued)

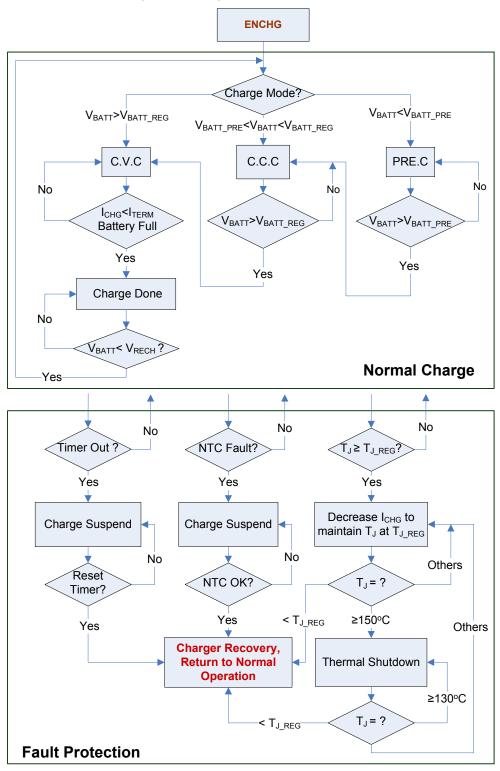


Figure 19: Charging Process

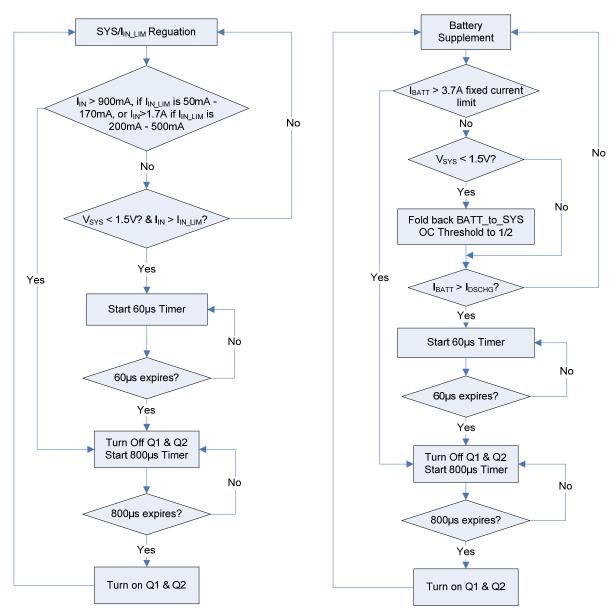


Figure 20: System Short-Circuit Protection

#### **APPLICATION INFORMATION**

## **Selecting a Resistor Divider for the NTC Sensor**

The NTC pin uses a resistor divider from the input source (VDD) to sense the battery temperature. The two resistors ( $R_{T1}$  and  $R_{T2}$ ) allow the high temperature limit and low temperature limit to be programmed independently (see Figure 21). The IC can fit most types of NTC resistors and different temperature operation range requirements with the two extra resistors.

For a given NTC thermistor, the  $R_{T1}$  and  $R_{T2}$  values depend on the type of NTC resistor used and can be calculated with Equation (1) and Equation (2):

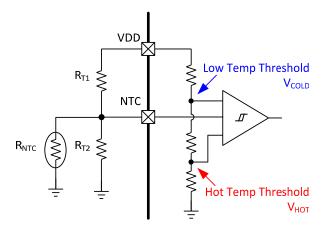


Figure 21: NTC Function Block

$$R_{T2} = \frac{\left(V_{\text{COLD}} - V_{\text{HOT}}\right) \times R_{\text{NTCH}} \times R_{\text{NTCL}}}{\left(V_{\text{HOT}} - V_{\text{COLD}}V_{\text{HOT}}\right) \times R_{\text{NTCL}} - \left(V_{\text{COLD}} - V_{\text{COLD}}V_{\text{HOT}}\right) \times R_{\text{NTCH}}}$$
(1)

$$R_{T1} = \frac{1 - V_{COLD}}{V_{COLD}} \times (R_{T2} // R_{NTCL})$$
 (2)

Where  $R_{\text{NTCH}}$  is the value of the NTC resistor at the high temperature of the required operating temperature range, and  $R_{\text{NTCL}}$  is the value of the NTC resistor at a low temperature.

For example, for a NCP18XH103 thermistor,  $R_{NTCL}$  is 27.219k $\Omega$  at 0°C, and  $R_{NTCH}$  is 4.161k $\Omega$  at 50°C. Using Equation (1) and Equation (2), calculate  $R_{T1}$  = 7.33k $\Omega$  and  $R_{T2}$  = 27.22k $\Omega$ , assuming that the NTC window is between 0°C and 50°C and using the  $V_{COLD}$  and  $V_{HOT}$  values from the EC table.

#### **Selecting the External Capacitor**

Like most low dropout regulators, the MP2662 requires external capacitors for regulator stability and voltage spike immunity. The device is specifically designed for portable applications requiring minimal board space and few components, so these capacitors must be selected correctly for optimal performance.

An input capacitor is required for stability. Connect a at least  $4.7\mu F$  ceramic capacitor (dielectric types X5R or X7R) between IN to GND for stable operation over the full load current range.

The IC is designed specifically to work with a very small ceramic output capacitor. A ceramic capacitor (dielectric types X5R or X7R) at least  $10\mu F$  is suitable for the application circuit. For the MP2662, the output capacitor should be connected between SYS and GND with thick traces and a small loop area.

A capacitor from BATT to GND is required. A at least  $4.7\mu F$  ceramic capacitor (dielectric types X5R or X7R) is suitable for the MP2662 application circuit.

A capacitor between VDD and GND is used to stabilize the VDD voltage to power the internal control and logic circuit. The typical value of this capacitor is 100nF.

#### **PCB Layout Guidelines**

Efficient PCB layout is critical for stable operation. For best results, follow the guidelines below.

- 1. Place the external capacitors as close to the IC as possible to ensure the smallest input inductance and ground impedance.
- 2. Place the PCB trace connecting the capacitor between VDD and GND very close to the IC.
- 3. Keep GND for the I<sup>2</sup>C wire clean.
- 4. Place the I<sup>2</sup>C wire in parallel.

## **TYPICAL APPLICATION CIRCUIT**

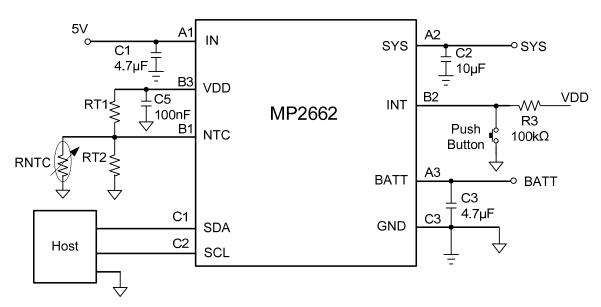


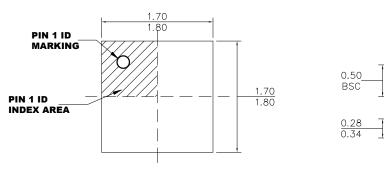
Figure 22: MP2662 Typical Application Circuit with 5V Input

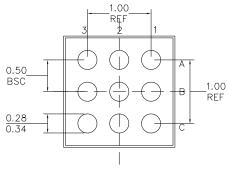
Table 5: Key BOM of Figure 22

Qty	Ref	Value	Description	Package	Manufacture
1	C1	4.7µF	Ceramic Capacitor, 25V, X5R or X7R	0603	Any
1	C3	4.7µF	Ceramic Capacitor, 16V, X5R or X7R	0603	Any
1	C2	10μF	Ceramic Capacitor, 16V, X5R or X7R	0603	Any
1	C5	100nF	Ceramic Capacitor, 16V, X5R or X7R	0603	Any

#### **PACKAGE INFORMATION**

## WLCSP-9 (1.75mmx1.75mm)





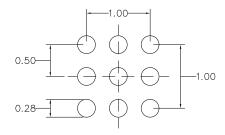
**TOP VIEW** 

**BOTTOM VIEW** 



#### **SIDE VIEW**

#### NOTE:



- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
  2) BALL COPLANARITY SHALL BE 0.05 MILLIMETER MAX.
- 3) JEDEC REFERENCE IS MO-211.
- 4) DRAWING IS NOT TO SCALE.

RECOMMENDED LAND PATTERN

## **REVISION HISTORY**

Revision#	Revision Date	Description	Pages updated
1.02	5/18/2020	1. Change "EV2662-C-00A" to EV2662-C-01A	Page 3
		Update the reset safety timer actions	Page 20
		Remove "UVLO or input over-voltage protection" from the INT section	Page 21
		4. Add CEB to the OTP map	Page 32
1.1	5/20/2021	Add "Safety-Related Certification" to Features section	Page 1
		2. Update pin description	Page 5
		Update EVKT content: Change "USB dongle" to "communication interface"	Page 3
		4. Update I <sup>2</sup> C interface description	Page 23-24
		5. Update register format	Page 7-32
		6. Improve the description of OPERATION	Page 17-24